INTEGRATED CIRCUITS

DATA SHEET



Monufacturer-user contract + educational material.

PCF8591 8-bit A/D and D/A converter

Product specification
Supersedes data of 2001 Dec 13

2003 Jan 27

Philips Semiconductors



PHILIPS

PCF8591

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PCF8591

1 FEATURES

- Single power supply
- Operating supply voltage 2.5 V to 6 V
- Low standby current
- Serial input/output via I²C-bus
- Address by 3 hardware address pins
- Sampling rate given by I²C-bus speed
- 4 analog inputs programmable as single-ended or differential inputs
- Auto-incremented channel selection
- \bullet Analog voltage range from V_{SS} to V_{DD}
- On-chip track and hold circuit
- 8-bit successive approximation A/D conversion
- Multiplying DAC with one analog output.

2 APPLICATIONS

- Closed loop control systems
- Low power converter for remote data acquisition
- Battery operated equipment
- Acquisition of analog values in automotive, audio and TV applications.

3 GENERAL DESCRIPTION

The PCF8591 is a single-chip, single-supply low power 8-bit CMOS data acquisition device with four analog inputs, one analog output and a serial I²C-bus interface. Three address pins A0, A1 and A2 are used for programming the hardware address, allowing the use of up to eight devices connected to the I²C-bus without additional hardware. Address, control and data to and from the device are transferred serially via the two-line bidirectional I²C-bus.

The functions of the device include analog input multiplexing, on-chip track and hold function, 8-bit analog-to-digital conversion and an 8-bit digital-to-analog conversion. The maximum conversion rate is given by the maximum speed of the I²C-bus.

Prose summary. Read it

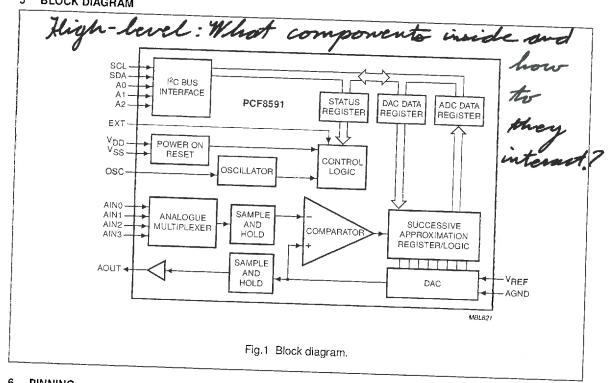
4 ORDERING INFORMATION

			O,
TYPE		PACKAGE	
NUMBER	NAME	DESCRIPTION	T
PCF8591P	DIP16	plastic dual in-line package; 16 leads (300 mil)	VERSION
PCF8591T	SO16	plastic small outline package; 16 leads; body width 7.5 mm	SOT38-4
		10 leads; body width 7.5 mm	SOT162-1

What designs would I use it in ?

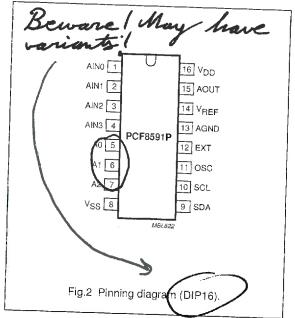
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BLOCK DIAGRAM

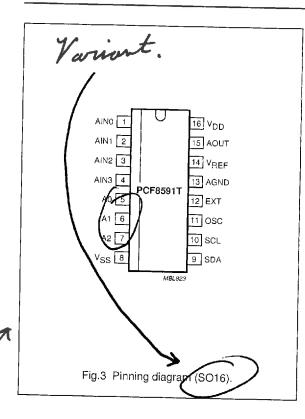


PINNING

SYMBOL	PIN	DESCRIPTION
AINO	1	analog inputs (A/D converter)
AIN1	2	5 Fate (FED CONVERTER)
AIN2	3	7
AIN3	4	7
A0	5	hardware address
A1	6	
A2	7	
V_{SS}	8	negative supply voltage
SDA	9	I ² C-bus data input/output
SCL	10	I ² C-bus clock input
OSC	11	oscillator input/output
EXT	12	external/internal switch for oscillator input
AGND	13	analog ground
V _{REF}	14	voltage reference input
AOUT	15	analog output (D/A converter)
V _{DD}	16	positive supply voltage



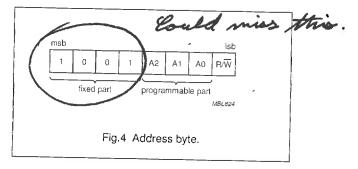
PCF8591



7 FUNCTIONAL DESCRIPTION

7.1 Addressing

Each PCF6591 device in an I²C-bus system is activated by sending a valid address to the device. The address consists of a fixed part and a programmable part. The programmable part must be set according to the address rins A0, A1 and A2. The address always has to be sent as the first byte after the start condition in the I²C-bus protocol. The last bit of the address byte is the read/write-bit which sets the direction of the following data transfer (see Figs 4, 16 and 17).



7.2 Control byte

The second byte sent to a PCF8591 device will be stored in its control register and is required to control the device function. The upper nibble of the control register is used for enabling the analog output, and for programming the analog inputs as single-ended or differential inputs. The lower nibble selects one of the analog input channels defined by the upper nibble (see Fig.5). If the auto-increment flag is set, the channel number is incremented automatically after each A/D conversion.

If the auto-increment mode is desired in applications where the internal oscillator is used, the analog output enable flag in the control byte (bit 6) should be set. This allows the internal oscillator to run continuously, thereby preventing conversion errors resulting from oscillator start-up delay. The analog output enable flag may be reset at other times to reduce quiescent power consumption.

The selection of a non-existing input channel results in the highest available channel number being allocated. Therefore, if the auto-increment flag is set, the next selected channel will be always channel 0. The most significant bits of both risk es are reserved for future functions and have to be set to logic 0. After a Power-on reset condition all bits of the confinit register are reset to logic 0. The D/A converter and the oscillator are disabled for power saving. The analog output is switched to a high-impedance state.

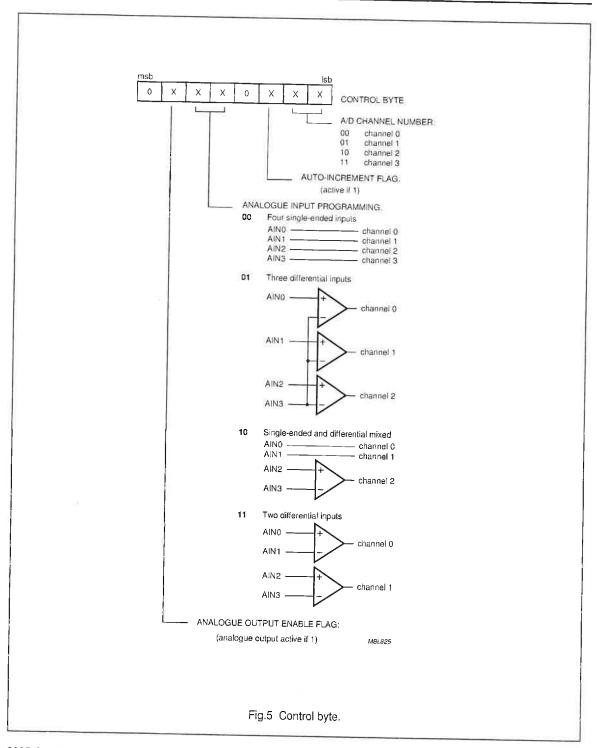
Sometimes # of components varies.

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7.3 D/A conversion

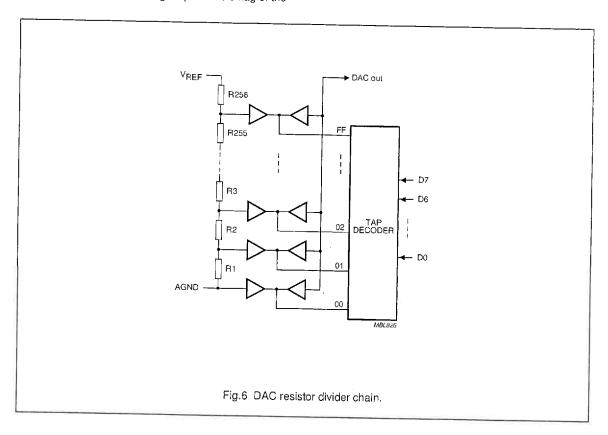
The third byte sent to a PCF8591 device is stored in the DAC data register and is converted to the corresponding analog voltage using the on-chip D/A converter. This D/A converter consists of a resister divider chain connected to the external reference voltage with 256 taps and selection switches. The tap-decoder switches one of these taps to the DAC output line (see Fig.6).

The analog output voltage is buffered by an auto-zeroed unity gain amplifier. This buffer amplifier may be switched on or off by setting the analog output enable flag of the

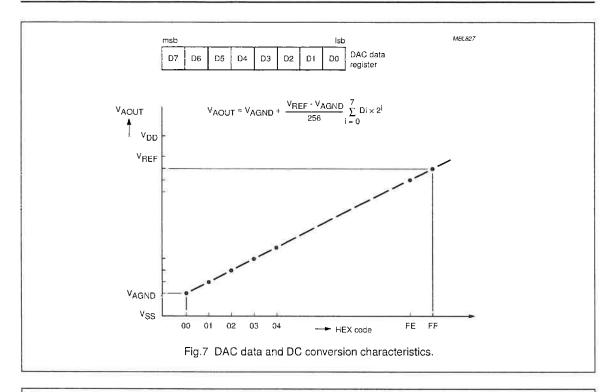
control register. In the active state the output voltage is held until a further data byte is sent.

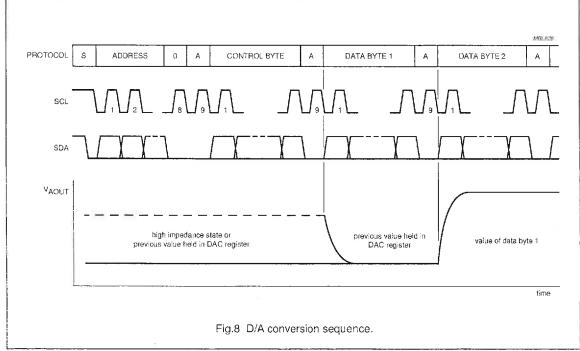
The on-chip D/A converter is also used for successive approximation A/D conversion. In order to release the DAC for an A/D conversion cycle the unity gain amplifier is equipped with a track and hold circuit. This circuit holds the output voltage while executing the A/D conversion.

The output voltage supplied to the analog output AOUT is given by the formula shown in Fig.7. The waveforms of a D/A conversion sequence are shown in Fig.8.



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7.4 A/D conversion

The A/D converter makes use of the successive approximation conversion technique. The on-chip D/A converter and a high-gain comparator are used temporarily during an A/D conversion cycle.

An A/D conversion cycle is always started after sending a valid read mode address to a PCF8591 device. The A/D conversion cycle is triggered at the trailing edge of the acknowledge clock pulse and is executed while transmitting the result of the previous conversion (see Fig.9).

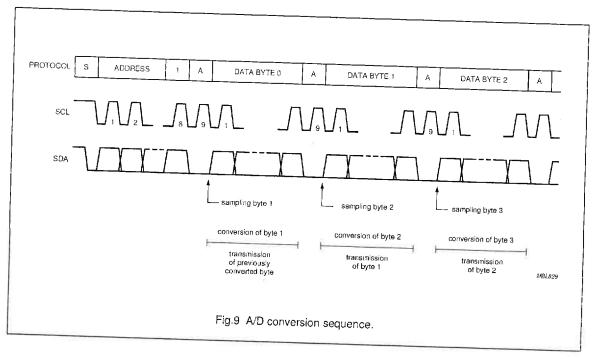
Once a conversion cycle is triggered an input voltage sample of the selected channel is stored on the chip and is

converted to the corresponding 8-bit binary code. Samples picked up from differential inputs are converted to an 8-bit twos complement code (see Figs 10 and 11).

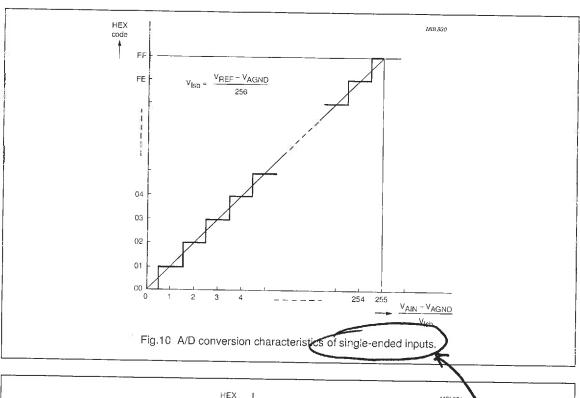
The conversion result is stored in the ADC data register and awaits transmission. If the auto-increment flag is set the next channel is selected.

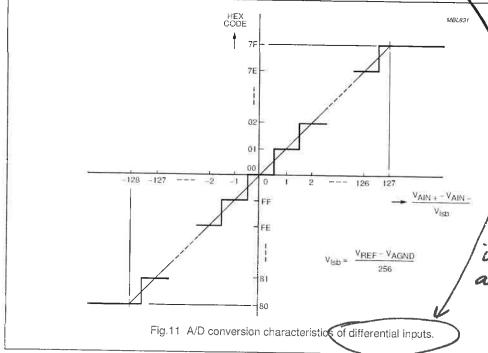
The first byte transmitted in a read cycle contains the conversion result code of the previous read cycle. After a Power-on reset condition the first byte read is a hexadecimal 80. The protocol of an I²C-bus read cycle is shown in Chapter 8, Figs 16 and 17.

The maximum A/D conversion rate is given by the actual speed of the I²C-bus.



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Which will you use? Don't understand took up & textools door online

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7.5 Reference voltage

For the D/A and A/D conversion either a stable external voltage reference or the supply voltage has to be applied to the resistor divider chain (pins V_{REF} and AGND). The AGND pin has to be connected to the system analog ground and may have a DC off-set with reference to V_{SS} .

A low frequency may be applied to the V_{REF} and AGND pins. This allows the use of the D/A converter as a one-quadrant multiplier; see Chapter 15 and Fig.7.

The A/D converter may also be used as a one or two quadrant analog divider. The analog input voltage is divided by the reference voltage. The result is converted to a binary code. In this application the user has to keep the reference voltage stable during the conversion cycle.

7.6 Oscillator

An on-chip oscillator generates the clock signal required for the A/D conversion cycle and for refreshing the auto-zeroed buffer amplifier. When using this oscillator the EXT pin has to be connected to V_{SS} . At the OSC pin the oscillator frequency is available.

If the EXT pin is connected to V_{DD} the oscillator output OSC is switched to a high-impedance state allowing the user to feed an external clock signal to OSC.

Still skimming.

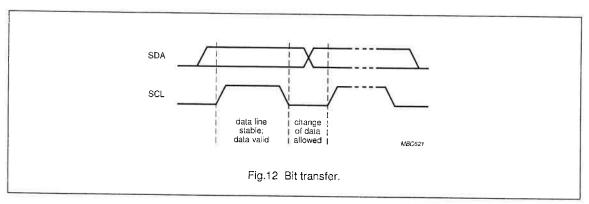
PCF8591

8 CHARACTERISTICS OF THE I2C-BUS

The I²C-bus is for bidirectional, two-line communication between different ICs or modules. The two lines are a serial data line (SDA) and a serial clock line (SCL). Both lines must be connected to a positive supply via a pull-up resistor. Data transfer may be initiated only when the bus is not busy.

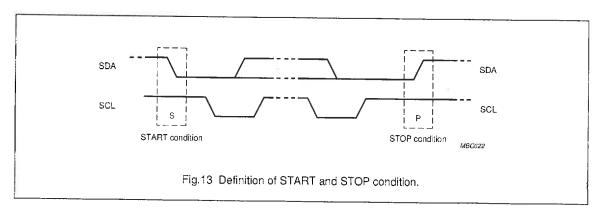
8.1 Bit transfer

One data bit is transferred during each clock pulse. The data on the SDA line must remain stable during the HIGH period of the clock pulse as changes in the data line at this time will be interpreted as a control signal.



8.2 Start and stop conditions

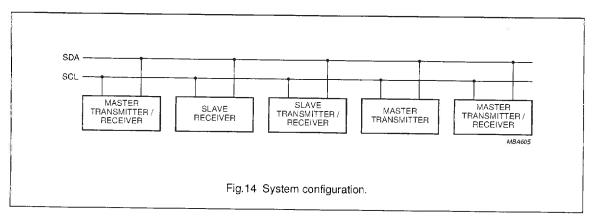
Both data and clock lines remain HIGH when the bus is not busy. A HIGH-to-LOW transition of the data line, while the clock is HIGH, is defined as the start condition (S). A LOW-to-HIGH transition of the data line while the clock is HIGH, is defined as the stop condition (P).



PCF8591

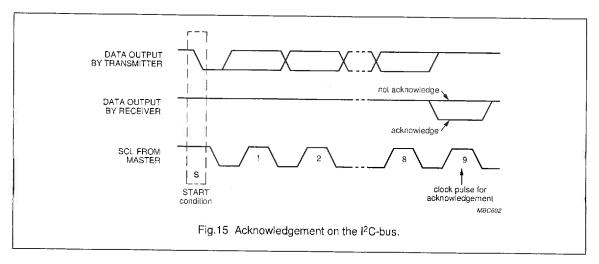
8.3 System configuration

A device generating a message is a 'transmitter', a device receiving a message is the 'receiver'. The device that controls the message is the 'master' and the devices which are controlled by the master are the 'slaves'.



8.4 Acknowledge

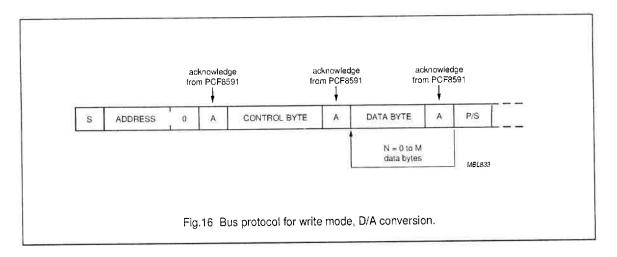
The number of data bytes transferred between the start and stop conditions from transmitter to receiver is not limited. Each data byte of eight bits is followed by one acknowledge bit. The acknowledge bit is a HIGH level put on the bus by the transmitter whereas the master also generates an extra acknowledge related clock pulse. A slave receiver which is addressed must generate an acknowledge after the reception of each byte. Also a master must generate an acknowledge after the reception of each byte that has been clocked out of the slave transmitter. The device that acknowledges has to pull down the SDA line during the acknowledge clock pulse, so that the SDA line is stable LOW during the HIGH period of the acknowledge related clock pulse. A master receiver must signal an end of data to the transmitter by **not** generating an acknowledge on the last byte that has been clocked out of the slave. In this event the transmitter must leave the data line HIGH to enable the master to generate a stop condition.

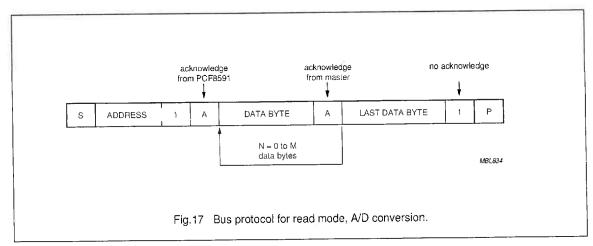


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8.5 I2C-bus protocol

After a start condition a valid hardware address has to be sent to a PCF8591 device. The read/write bit defines the direction of the following single or multiple byte data transfer. For the format and the timing of the start condition (S), the stop condition (P) and the acknowledge bit (A) refer to the I²C-bus characteristics. In the write mode a data transfer is terminated by sending either a stop condition or the start condition of the next data transfer.





On 10 HANDLING

8-bit A/D and D/A converter

PCF8591

LIMITING accordance	with the Absolute Maximum Rating System (IEC	60134).	1 m	omer
SYMBOL	PARAMETER	MIN.	Not A	UNIT
)	supply voltage (pin 16)	-0.5	4 8.0	V
	input voltage (any input)	-0.5	$V_{DD} + 0.5$	V
	DC input current	-	±10	mA
	DC output current	_	±20	mA
o, Iss	V _{DD} or V _{SS} current	_	±50	mA-
1	total power dissipation per package	- /	300	mW
	power dissipation per output	-	100	mW
mb dr	operating ambient temperature	-40	+85	R
slo	storage temperature	-65	+150	°C

Inputs and outputs are protected against electrostatic discharge in normal handling. However it is good practice to take normal precautions appropriate to handling MOS devices (see "Handling MOS devices").

Break these rules and manufacturer is in compliance even if part explades.

Don't use restimation.

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Super useful for power lot, but consider variation,

8-bit A/D and D/A converter

PCF8591

11 DC CHARACTERISTICS

V_{DD} = 2.5 V to 6 V; V_{SS} = 0 V; T_{amb} = $\sqrt{40}$ °C to +85 °C unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Supply					<u> </u>	1
V _{DD}	supply voltage (operating)		2.5	_	6.0	٧
I _{DD}	emply current					
y	standby	$V_l = V_{SS}$ or V_{DD} ; no load	_	1	15	μΑ
(operating, AOUT off	f _{SCL} = 100 kHz	_	125	250	μΑ
	operating, AOUT active	f _{SCL} = 100 kHz	_	0.45	1.0	mA
V _{POR}	Power-on reset level	note 1	0.8	_	2.0	V
Digital inp	outs/output: SCL, SDA, A0, A1,	A2				
VIL	LOW level input voltage		0	-	$0.3 \times V_{DD}$	V
V _{IH}	HIGH level input voltage		$0.7 \times V_{DD}$	_	V _{DD}	٧
l _L	leakage current					
	A0, A1, A2	$V_I = V_{SS}$ to V_{DD}	-250	_	+250	nA
	SCL, SDA	$V_I = V_{SS}$ to V_{DD}	-1	_	+1	μΑ
Ci	input capacitance		_	-	5	рF
loL	LOW level SDA output current	V _{OL} = 0.4 V	3.0	_	_	mΑ
Reference	voltage inputs					,
V_{REF}	reference voltage	V _{REF} > V _{AGND} ; note 2	V _{SS} + 1.6	_	V _{DD}	V
V_{AGND}	analog ground voltage	V _{REF} > V _{AGND} ; note 2	V _{SS}	_	V _{DD} - 0.8	٧
ILI	input leakage current		-250	_	+250	nA
R _{REF}	input resistance	pins V _{REF} and AGND	_	100	_	kΩ
Oscillator	: OSC, EXT					•
l _{LI}	input leakage current		-	_	250	nA
fosc	oscillator frequency		0.75	_	1.25	MHz

Notes

- 1. The power on reset circuit resets the I^2C -bus logic when V_{DD} is less than V_{POR} .
- 2. A further extension of the range is possible, if the following conditions are fulfilled:

$$\frac{V_{REF} + V_{AGND}}{2} \ge 0.8 \text{ V}, V_{DD} - \frac{V_{REF} + V_{AGND}}{2} \ge 0.4 \text{ V}$$

Lawyers: The big print gives. The fine print takes away.

Engineers: The big print takes away.

The fine print gives.

2003 Jan 27

Why?

Marininges complexity.

Minimiyes complexity

PCF8591

12 D/A CHARACTERISTICS

 V_{DD} = 5.0 V; V_{SS} = 0 V; V_{REF} = 5.0 V; V_{AGND} = 0 V; R_L = 10 k Ω ; C_L = 100 pF; T_{amb} = -40 °C to +85 °C unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Analog ou	utput					1
V _{OA}	output voltage	no resistive load	V _{SS}	_	V _{DD}	V
	<u></u>	$R_L = 10 \text{ k}\Omega$	V _{SS}	-	$0.9 \times V_{DD}$	٧
I _{LO}	output leakage current	AOUT disabled	-	_	250	nA
Accuracy				· · · · · · · · · · · · · · · · · · ·		
OS _e	offset error	T _{amb} = 25 °C	_	_	50	mV
L _e	linearity error		_	_	±1.5	LSB
G _e	gain error	no resistive load	-	-	1	%
t _{DAC}	settling time	to ½ LSB full scale step	-	1-	90	μs
f _{DAC}	conversion rate		_	-	11.1	kHz
SNRR	supply noise rejection ratio	f = 100 Hz; V _{DDN} = 0.1 × V _{PP}	_	40	_	dB

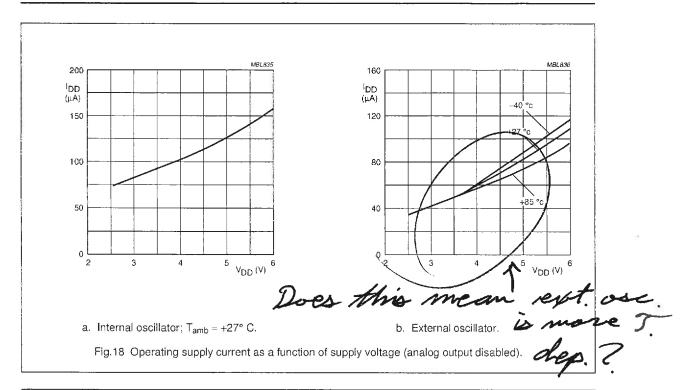
13 A/D CHARACTERISTICS

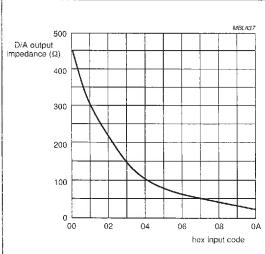
 $V_{DD} = 5.0 \text{ V; } V_{SS} = 0 \text{ V; } V_{REF} = 5.0 \text{ V; } V_{AGND} = 0 \text{ V; } R_S = 10 \text{ k}\Omega; \\ T_{amb} = -40 \text{ °C to } +85 \text{ °C unless otherwise specified.}$

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Analog in	puts			<u> </u>		
V _{IA}	analog input voltage		V _{SS}	8_	V _{DD}	V
I _{LIA}	analog input leakage current		-	_	100	nA
CIA	analog input capacitance			10		pF
CID	differential input capacitance		-	10	_	pF
V _{IS}	single-ended voltage	measuring range	V _{AGND}	-	V _{REF}	V
V _{ID}	differential voltage	measuring range; V _{FS} = V _{REF} - V _{AGND}	-V _{FS} 2	_	+V _{FS} 2	V
Accuracy						
OS _e	offset error	T _{amb} = 25 °C	_	1-	20	mV
Le	linearity error		_	_	±1.5	LSB
Ge	gain error		_	_	1	%
GS _e	small-signal gain error	$\Delta V_i = 16 LSB$	_	_	5	%
CMRR	common-mode rejection ratio		_	60	_	dB
SNRR	supply noise rejection ratio	f = 100 Hz; V _{DDN} = 0.1 × V _{PP}	-	40	-	dB
t _{ADC}	conversion time		_	1-	90	μs
f _{ADC}	sampling/conversion rate		_	-	11.1	kHz

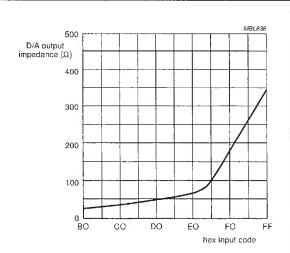
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a. Output impedance near negative power rail; $T_{amb} = +27~^{\circ}C. \label{eq:Tamb}$



b. Output impedance near positive power rail; $T_{amb} = +27 \, ^{\circ}\text{C}$.

The x-axis represents the hex input-code equivalent of the output voltage.

Fig. 19 Output impedance of analog output buffer (near power rails).

PCF8591

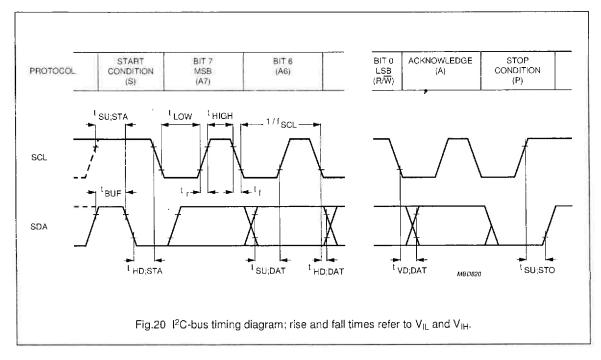
14 AC CHARACTERISTICS

All timing values are valid within the operating supply voltage and ambient temperature range and reference to V_{IL} and V_{IH} with an input voltage swing of V_{SS} to V_{DD} .

SYMBOL	PARAMETER	MIN.	TYP.	MAX.	UNIT
l ² C-bus timin	g (see Fig.20; note 1)			- ·	
f _{SCL}	SCL clock frequency	_	-	100	kHz
t _{SP}	tolerable spike width on bus	_	<u> </u>	100	ns
t _{BUF}	bus free time	4.7	_	-	μs
t _{SU;STA}	START condition set-up time	4.7	-	-	μs
thd;STA	START condition hold time	4.0	_	-	μs
t _{LOW}	SCL LOW time	4.7	_	-	μs
t _{HIGH}	SCL HIGH time	4.0	-	-	μs
t _r	SCL and SDA rise time	_	_	1.0	μs
t _f	SCL and SDA fall time	1 -	-	0.3	μs
t _{su;dat}	data set-up time	250	-		ns
t _{HD;DAT}	data hold time	0	-	_	ns
t _{VD:DAT}	SCL LOW-to-data out valid	_	-	3.4	μs
t _{SU;STO}	STOP condition set-up time	4.0	-	_	μs

Note

1. A detailed description of the I²C-bus specification, with applications, is given in brochure *"The I²C-bus and how to use it"*. This brochure may be ordered using the code 9398 393 40011.



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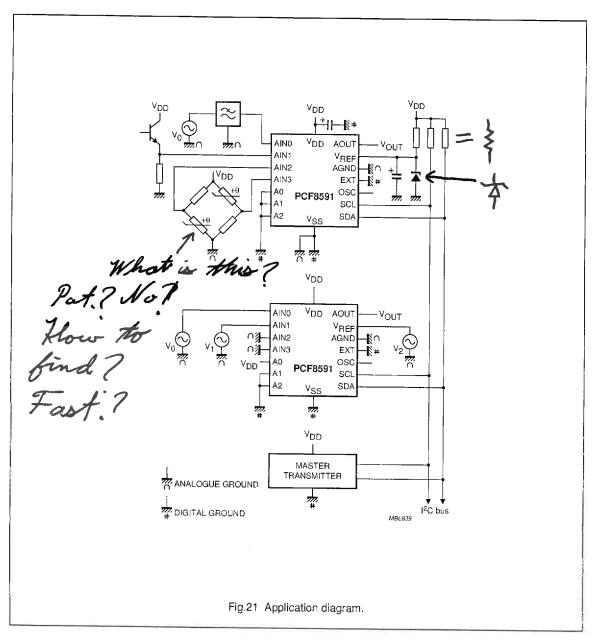
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15 APPLICATION INFORMATION

Inputs must be connected to V_{SS} or V_{DD} when not in use. Analog inputs may also be connected to AGND or V_{REF}.

In order to prevent excessive ground and supply noise and to minimize cross-talk of the digital to analog signal paths the user has to design the printed-circuit board layout very carefully. Supply lines common to a PCF8591 device and noisy digital circuits and ground loops should be avoided. Decoupling capacitors (>10 μ F) are recommended for power supply and reference voltage inputs.

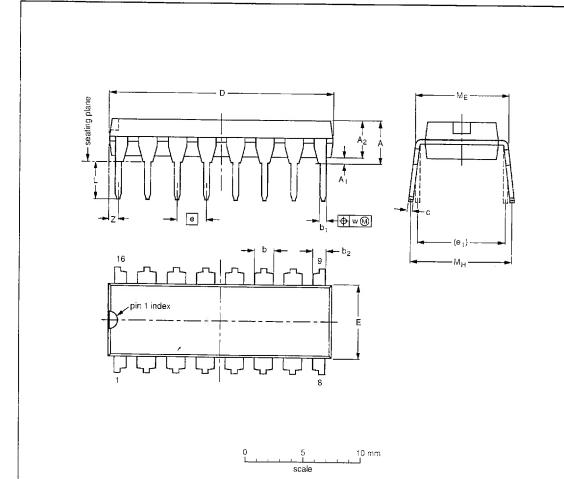


PCF8591

16 PACKAGE OUTLINES

DIP16: plastic dual in-line package; 16 leads (300 mil)

SOT38-4



DIMENSIONS (inch dimensions are derived from the original mm dimensions)

					w 11 0 111 L				,							
UNIT	A max.	A 1 mln,	A ₂ max.	ь	b ₁	b ₂	С	D (1)	E ⁽¹⁾	е	e ₁	L	ME	Мн	w	Z ⁽¹⁾ max.
mm	4.2	0.51	3.2	1.73 1.30	0.53 0.38	1.25 0.85	0.36 0.23	19.50 18.55	6.48 6.20	2.54	7.62	3.60 3.05	8.25 7.80	10.0 8,3	0.254	0.76
inches	0.17	0.020	0.13	0.068 0.051	0.021 0.015	0.049 0.033	0.014 0.009	0.77 0.73	0.26 0.24	0.10	0.30	0.14 0.12	0.32 0.31	0.39 0.33	0.01	0.030

Note

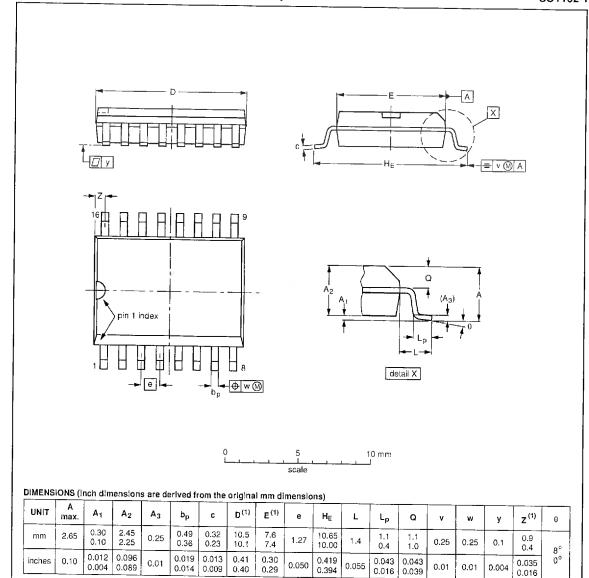
^{1.} Plastic or metal protrusions of 0.25 mm maximum per side are not included.

OUTLINE		REFER	ENCES		EUROPEAN	ılı .
VERSION	IEC	JEDEC	EiAJ)	PROJECTION	ISSUE DATE
SOT38-4		<u> </u>			€	92-11-17- 95-01-14

PCF8591



SOT162-1



Note

^{1.} Plastic or metal protrusions of 0.15 mm maximum per side are not included.

OUTLINE		REFERI	ENCES	 EUROPEAN	
VERSION	IEC	JEDEC	ElAJ	PROJECTION	ISSUE DATE
SOT162-1	075E03	MS-013		€	-97-05-22 99-12-27

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17 SOLDERING

17.1 Introduction to soldering through-hole mount packages

This text gives a brief insight to wave, dip and manual soldering. A more in-depth account of soldering ICs can be found in our "Data Handbook IC26; Integrated Circuit Packages" (document order number 9398 652 90011).

Wave soldering is the preferred method for mounting of through-hole mount IC packages on a printed-circuit board.

17.2 Soldering by dipping or by solder wave

The maximum permissible temperature of the solder is 260 °C; solder at this temperature must not be in contact with the joints for more than 5 seconds.

The total contact time of successive solder waves must not exceed 5 seconds.

The device may be mounted up to the seating plane, but the temperature of the plastic body must not exceed the specified maximum storage temperature ($T_{sig(max)}$). If the printed-circuit board has been pre-heated, forced cooling may be necessary immediately after soldering to keep the temperature within the permissible limit.

17.3 Manual soldering

Apply the soldering iron (24 V or less) to the lead(s) of the package, either below the seating plane or not more than 2 mm above it. If the temperature of the soldering iron bit is less than 300 °C it may remain in contact for up to 10 seconds. If the bit temperature is between 300 and 400 °C, contact may be up to 5 seconds.

17.4 Suitability of through-hole mount IC packages for dipping and wave soldering methods

PACKAGE	SOLDE	RING METHOD
	DIPPING	WAVE
DBS, DIP, HDIP, SDIP, SIL	suitable	suitable ⁽¹⁾

Note

1. For SDIP packages, the longitudinal axis must be parallel to the transport direction of the printed-circuit board.

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LEVEL	DATA SHEET STATUS(1)	PRODUCT STATUS(2)(3)	DEFINITION
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Contact information

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